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INCREASING SECONDARY DATA OR ENHANCED ADAPTIVE DISPLACED PHASE CENTERED ANTENNA APPROACHES

Capraro Technologies, Inc.

Gerard T. Capraro, Christopher T. Capraro, and Donald D. Weiner

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1. INTRODUCTION

In this chapter, some partially adaptive space-time processing (PSTAP) schemes [1] are compared according to the performance obtained using a measured radar data set called multichannel airborne measurements (MCARM). Algorithms studied include adaptive displaced phase centered antenna (ADPCA), factored post Doppler (FTS), extended factored approach (EFA), joint-domain localized (JDL), subarraying ADPCA, subarraying EFA, subarraying FTS, and beamspace ADPCA. We use normalized test statistics which achieve constant false alarm probability in homogenous clutter. The performance of the algorithms is studied in section 2.

1.1 Subarraying ADPCA

Let N be the number of antenna elements, M be the number of pulses per coherent processing interval (CPI), K_s be the number of beams formed and, K_t be the number of pulses which will be adaptively processed.

First the pre-processing performs the beamforming and forms groups of \mathbf{K}_t pulses from \mathbf{K}_s beam outputs as

$$\tilde{\mathbf{X}}_{k}(\mathbf{p}) = (\mathbf{A}_{\mathbf{p}} \otimes \mathbf{G})^{H} \mathbf{X}_{k} \quad ; \quad \mathbf{p} = 0, 1, 2, \dots, P - 1$$
 (1)

where X_k is the space-time snapshot organized as

$$\mathbf{X}_{k} = \left[\mathbf{x}_{1,1,k}, \mathbf{x}_{2,1,k}, \cdots, \mathbf{x}_{N,1,k}, \mathbf{x}_{1,2,k}, \cdots \mathbf{x}_{N,M,k} \right]^{\mathsf{T}}$$
(2)

where $\mathbf{x_{i,j,k}}$ is the observation from the i^{th} antenna element, j^{th} pluse and k^{th} range cell [2], \mathbf{G} is the $\mathbf{N} \times \mathbf{K_s}$ beamformer matrix, and

$$\mathbf{A}_{\mathbf{p}} = \begin{bmatrix} \mathbf{0}_{\mathbf{p} \times \mathbf{K}_{t}} \\ \mathbf{I}_{\mathbf{K}_{t}} \\ \mathbf{0}_{(\mathbf{M} - \mathbf{K}_{t} - \mathbf{p}) \times \mathbf{K}_{t}} \end{bmatrix}$$
(3)

where the notation $0_{q\times m}$ refers to an $q\times m$ matrix of zeros and I_{Kt} is a $K_t\times K_t$ identity matrix. It is possible to form sub-CPIs which overlap by any number of pulses as in [2]. In this report we focus on the case where K_t is set to 3 and consecutive sub-CPIs overlap two pulses. We also focus on $K_s=3$ in this report.

The beamformer matrix mentioned in [1] (See equation (256) on page 14 in [1].) is used in our processing. This matrix is

$$G = \begin{bmatrix} g_0 & 0 & 0 \\ g_1 & g_0 \\ \vdots & g_1 & \ddots \\ g_{N'-1} & \vdots & g_0 \\ & g_{N'-1} & \ddots & g_1 \\ & & & \vdots \\ 0 & 0 & g_{N'-1} \end{bmatrix}$$

$$(4)$$

where $\mathbf{g} = (\mathbf{g_0}, \mathbf{g_1}, \mathbf{g_2}, \dots, \mathbf{g_{N'-1}})^T$ is a vector of scalar components and $\mathbf{N'} = \mathbf{N} \cdot \mathbf{K_s} + 1$. The vector \mathbf{g} could be defined as any of the popular windows or tapering from the DSP [3] or radar [4] literature. Here we take \mathbf{g} as either a uniform window (all ones) or the hamming window in our tests. The uniform window gave better results than the hamming window so we present results for the uniform window. The \mathbf{G} matrix in (4) represents a case where the beams use identical beamforming using a length $\mathbf{N'}$ element subaperture, with each subaperture shifted by one element from the previous one. The resulting processing is sometimes called subarraying.

Each vector produced by the pre-processing (1) will be adaptively processed as

$$\mathbf{y}_{k}(\mathbf{p}) = \mathbf{S}^{H}\mathbf{R}_{k}^{-1}(\mathbf{p})\tilde{\mathbf{X}}_{k}(\mathbf{p})/\Phi$$
 (5)

where

$$\mathbf{S} = (\mathbf{I}_{\mathbf{K}\mathbf{t}} \otimes \mathbf{G})^{\mathbf{H}} \mathbf{S}_{\mathbf{e}} \tag{6}$$

with Se as

$$S_{e} = S_{t} \otimes S_{s} \tag{7}$$

where S_s is the N×1 spatial target vector [1] and S_t is a K_t ×1 vector of binomial coefficients of altering sign normally used in ADPCA [2]. The term Φ in the denominator of (5) is the normalization to provide CFAR in homogeneous clutter and is given by

$$\Phi = \mathbf{S}^{\mathsf{H}} \mathbf{R}_{\mathsf{k}}^{-1} \mathbf{S} \tag{8}$$

 $\mathbf{R}_{\mathbf{k}}$ is the interference-plus-noise covariance matrix estimated from the pre-processed data from the \mathbf{Q} adjacent range cells (reference data), excluding the cell- under-test and the two closest cells on each side of the cell-under-test, as follows

$$R_{k}(p) = \frac{1}{Q} \sum_{i=k-Q/2-2, i\neq k-2, k-1, k, k+1, k+2}^{k+Q/2+2} \tilde{X}_{i}(p) \tilde{X}_{i}(p) H$$
 (9)

After the adaptive processing, post processing is performed which can be expressed as

$$\tilde{\mathbf{Y}}_{k} = \left[\mathbf{y}_{k}(\mathbf{0}), \mathbf{y}_{k}(\mathbf{1}), \dots, \mathbf{y}_{k}(\mathbf{P} - \mathbf{1}) \right]^{\mathsf{T}}$$
(10)

$$\tilde{\mathbf{Z}}_{\mathbf{k},\mathbf{m}} = \mathbf{f}_{\mathbf{m}}^{\mathsf{H}} \, \tilde{\mathbf{Y}_{\mathbf{k}}} \tag{11}$$

assuming the target is in the m^{th} Doppler bin. In this case f_m is the m^{th} column of a $P \times P$ Doppler filter matrix F, and $Z_{k,m}$ is the final output whose magnitude will be compared to a threshold to produce a decision. In the case of plain ADPCA (not subarraying) we set $G = I_N$.

1.2 Subarraying FTS (BFTS)

In subarraying factored post-Doppler (FTS) STAP, the pre-processing is described by (1) with $A_p = f_p$ and P = M. f_p is the p^{th} column of an $M \times M$ Doppler filter matrix $F = [f_0, f_1, f_2,, f_{M-1}]$. G is the $N \times K_s$ beamforming matrix as in subarraying ADPCA. This

pre-processing transforms the signal into Doppler space. To test for a target at a particular Doppler frequency, only one pre-processing from (1) needs to be calculated. In FTS a single pulse at a time is processed (\mathbf{K}_t =1), and \mathbf{p} indicates the Doppler bin in question. The adaptive processing in (5) is employed with the steering vector \mathbf{S} defined as in (6) with \mathbf{K}_t =1. Here \mathbf{S}_e is set to \mathbf{S}_s in (7), since \mathbf{K}_t is 1. Post processing is not usually employed, so $|\mathbf{y}_k|(\mathbf{p})|$ is compared to a threshold to test for a target in the \mathbf{p}^{th} Doppler bin. In the case of plain FTS we set $\mathbf{G} = \mathbf{I}_N$.

1.3 Subarraying EFA (BEFA)

The subarraying extended factored approach (BEFA) is a slight extension of the subarraying FTS approach. In BEFA, adaptive processing is applied to several adjacent Doppler bins instead of just one. In our subarraying EFA implementation, the scheme adapts over 3 adjacent bins. The pre-processing is as in (1) with $A_p=[f_{p-1}, f_p, f_{p+1}]$. The pre-processing performs both transformation and selection. The adaptive processing is as in subarraying ADPCA with $S_t=[0\ 1\ 0]^T$. Post processing is not employed. In the case of plain EFA we set $G=I_N$.

1.4 Beamspace ADPCA (BeamAD)

In beamspace ADPCA, the pre-processing can be defined as in (1) with A_p as in (3), and $G=[f_{N,1}, f_{N,2}, ..., f_{N,Ks}]$, where $f_{N,j}$ $j=1, ..., K_s$ are the columns of N×N DFT matrix corresponding to K_s angle bins. For simplicity assume $K_s=3$ and the target to be detected is in the 2^{nd} angle bin. The adaptive processing is as in subarraying ADPCA with $S=[0 \ 1 \ 0] \otimes S_t$, where S_t is a $K_t \times 1$ vector of binomial coefficients of altering sign as in subarraying ADPCA. In our tests K_s and K_t were set to 3. The post processing is as in subarraying ADPCA as described by (10) and (11).

1.5 Joint-domaint localized approach (JDL)

In JDL, the pre-processor performs two dimensional transformation and selection. The data is transformed from the space-time domain into the angle-Doppler domain. This pre-processing can be described as in (1), with $A_p=[f_{m,1}, f_{m,2}, ..., f_{m,Kt}]$, where $f_{m,j}$ $j=1,, K_t$ are the columns of an M×M DFT matrix corresponding to the K_t Doppler bins and with $G=[f_{n,1}, f_{n,2}, f_{n,2},$

..., f_{n,K_s}], where $f_{n,j}$ j=1, ..., K_s are the columns of N×N DFT matrix corresponding to the K_s angle bins. As for FTS and EFA only the post processing corresponding to a single p must be calculated to test for a target at a particular Doppler and Angular frequency.

If we again focus on $K_t = K_s = 3$ and consider the case where the target to be detected is in 2^{nd} angle and the 2^{nd} Doppler bin then, the adaptive processing can be described as in (5) with

$$S = [0 \ 1 \ 0] \otimes [0 \ 1 \ 0] \tag{12}$$

More precisely, the steering vector has all its entries equal to zero except for the one corresponding to the angle and Doppler bin of the target. No post-processing is employed for JDL.

2. REAL DATA PERFORMANCE

To test the STAP algorithms described in section 1, we use data that comes from the MCARM database flight 5 acquisition 575. See [5] and [6] for detailed information about the MCARM program and the data. For each experiment, a single target signal with amplitude 0.05, a particular normalized Doppler frequency and a particular spatial frequency was inserted in a particular range bin. Reference data are selected from consecutive cells on each side of the cell-under-test (see Appendix for complete details on all schemes). We employ normalized test statistics, which provide a constant false alarm rate (CFAR) characteristic for homogenous clutter [7].

For each example, we provide plots of the magnitude of the normalized test statistics for a set of ranges including the target range. We judge a scheme by how large the test statistic is at the target range in comparison to other ranges.

In the first example, we inserted a target at range bin 150 for the cases shown in Table 1. The location of the targets and an estimate of the clutter (plus noise) power spectral density (psd) are illustrated in Fig. 1. As visible from Fig. 1, the psd estimate used is rather crude and is provided to give a rough description of the clutter environment. In the estimate no neighboring range cells are averaged and a Blackman window is used. Some of the artifacts can be removed by averaging, but this was not considered necessary in this case.

Case	Normalized D	oppler Normalized Spatial
	Frequency	Frequency
a	-0.2656	-0.2656
ь	-0.0312	-0.0312
c	0.125 0.203	
đ	-0.0312	-0.1875
e	-0.0312	0.203
f	0.3593	-0.1875
g	-0.1875	0.3593

Table 1) Test cases for each example.

A summary of the results is given in Table 2. Fig. 2 through Fig. 8 presents the results for most of the schemes tested. FTS and subarraying FTS generally perform poorly, so their results are not shown. In this example, generally JDL provides best results. JDL is best in every case except cases d and e. In case d, BeamAD slightly outperforms JDL and ADPCA, but the difference is quite small. In case e, EFA outperforms the other schemes, however JDL also performs well. The apparently large interference indicated by Fig. 1 near the normalized Doppler frequency -0.0312 could be the reason why JDL is not best in these two cases. These are apparently difficult cases where no scheme can really excel. Table 2 shows that case b is also quite difficult, but here JDL performs much better than the other schemes. The extra clutter ridges in Fig. 1 are discussed in [8] and [9].

In the second example, we inserted a target at range bin 350. We present results for the same cases in Table 1. The best three schemes for all the cases are given in Table 3. The normalized test statistics for the six best schemes tested for each case in Table 1 are given in Figures 10 through 16. The results indicate that none of the schemes always outperforms all the others. However, post-Doppler algorithms are generally better than pre-Doppler algorithms. Either JDL or EFA were best in all but case b. In case b, where the target is inserted in the largest clutter of all cases, BEFA is only slightly better than EFA and JDL.

Next, we inserted a target at range bin 415. We present results for the same cases as in previous examples. The location of the targets and an estimate of the clutter psd is given in Fig.

17. A summary of the results is provided in Table 4. The normalized test statistics are given in Fig. 18 through Fig. 24. Here for cases b, c, d, f, and g either JDL, or BEFA provide best performance. In the other cases, EFA is best and JDL also performs well. JDL performs well in every case except case b. Even in this case, its performance is not bad. EFA and its beamspace version are best in some cases and near-best in others. ADPCA and its beamspace version give good performance in number of cases, but these schemes were never best in this example.

Finally, we test the same cases as before when the target is inserted in range bin 500. The locations of the targets and an estimate of the clutter psd is given in Fig. 18. A summary of the results is given in Table 5. Fig. 26 through Fig. 32 provide the test statistics for the six best schemes in each case. Again the beamspace post-doppler algorithms, JDL and BEFA, outperform the element space algorithms except for case e where ADPCA is best.

4. CONCLUSIONS

In our tests, JDL and EFA generally perform very well. Subarrayed EFA also shows good performance in many situations. The common element these schemes share is post doppler processing. This type of processing, used in the correct way, appears to be superior in these real data cases. This appears to be the major result of this study. To a much lesser extent, schemes which estimate fewer parameters (lower degree of freedom (DOF) schemes) do appear to provide some benefits in non-homogeneous cases. The number of parameters estimated for each scheme is listed in Appendix 1. Some further discussion and analysis of non-homogeneous cases is given in Appendix 2. When there is a strong interference near the target, as in case d and e in Table 5 for example, ADPCA performs well and sometimes outperforms all the other schemes. The reason appears to be related to the extra whitening provided by its steering vector.

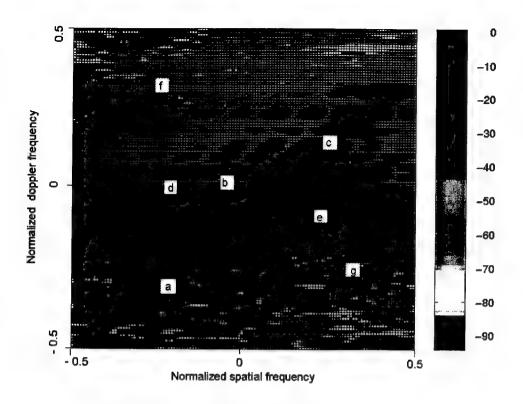


Figure 1) Power spectrum plot of range 150

Case	Normalized Doppler Frequency	Normalized Spatial Frequency	The 3 best schemes	D	Figure
а	-0.2656	-0.2656	JDL BEFA BeamAD	36 25 16	2
b	-0.0312	-0.0312	JDL	10	3
С	0.125	0.203	JDL EFA	30 12	4
d	-0.0312	-0.1875	BeamAD JDL ADPCA	8 7 6	5
е	-0.0312	0.203	EFA JDL BeamAD	20 12 5	6
f	0.3593	-0.1875	JDL EFA BeamAD	43 38 30	7
g	-0.1875	0.3593	JDL BEFA BeamAD	38 18 7	8

Table 2) The three best schemes for all the cases when the target is inserted at range 150. (D is the approximate difference between the normalized test statistic at the target and the largest peak in the normalized test statistic at some other range)

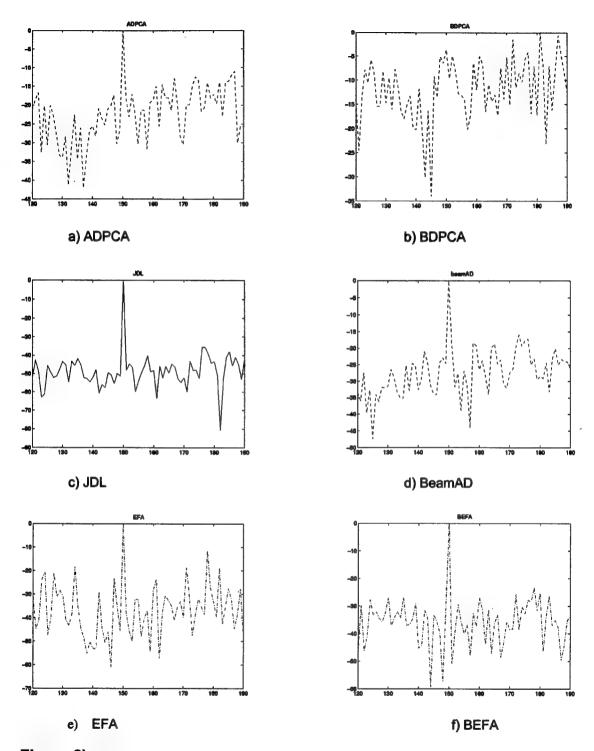


Figure 2) Performance comparison when target is inserted at range150, case a.

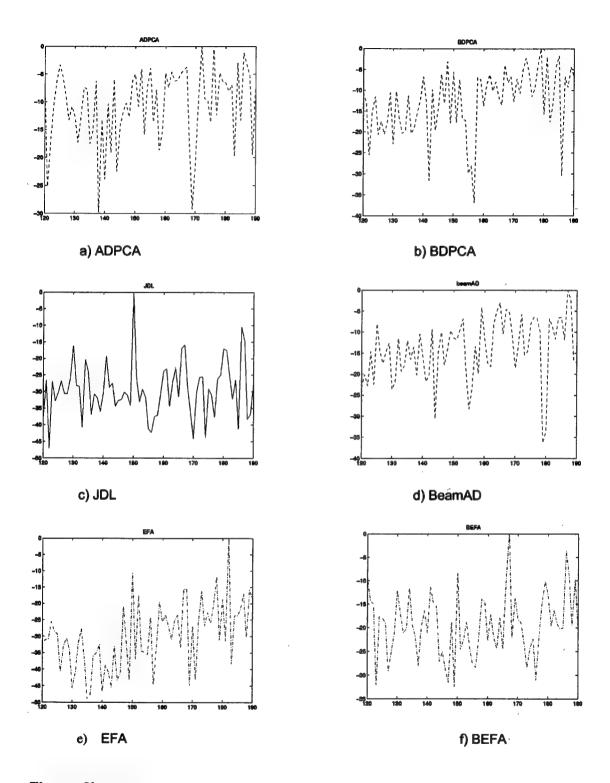


Figure 3) Performance comparison when target is inserted at range 150, case b.

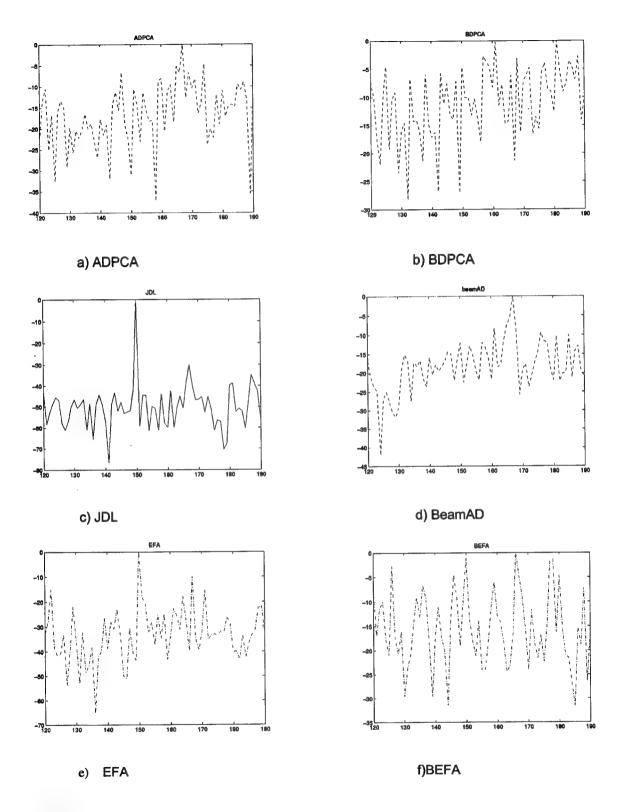


Figure4) Performance comparison when target is inserted at range 150, case c.

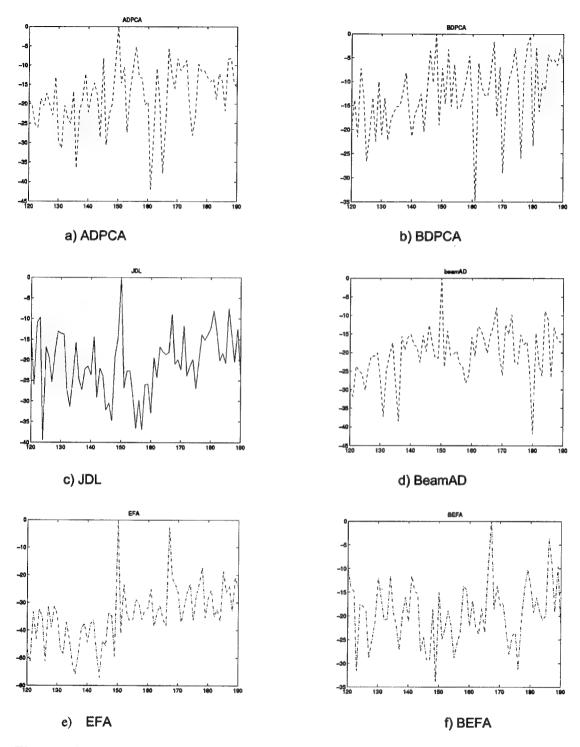


Figure 5) Performance comparison when target is inserted at range 150, case d.

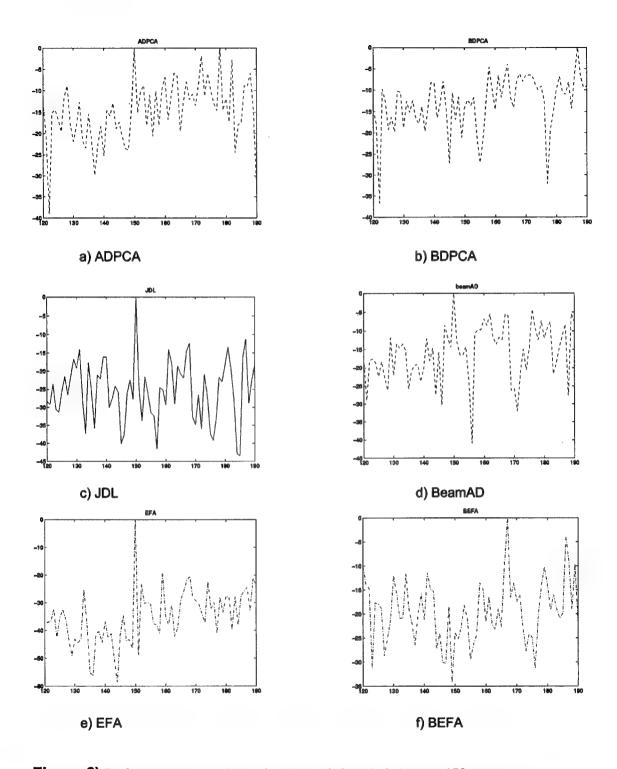


Figure 6) Performance comparison when target is inserted at range 150, case e.

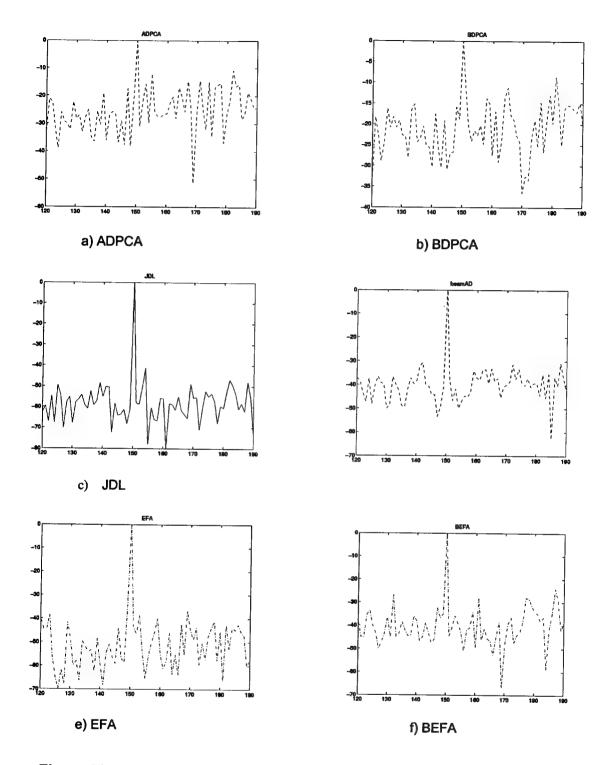


Figure 7) Performance comparison when target is inserted at range 150, case f.

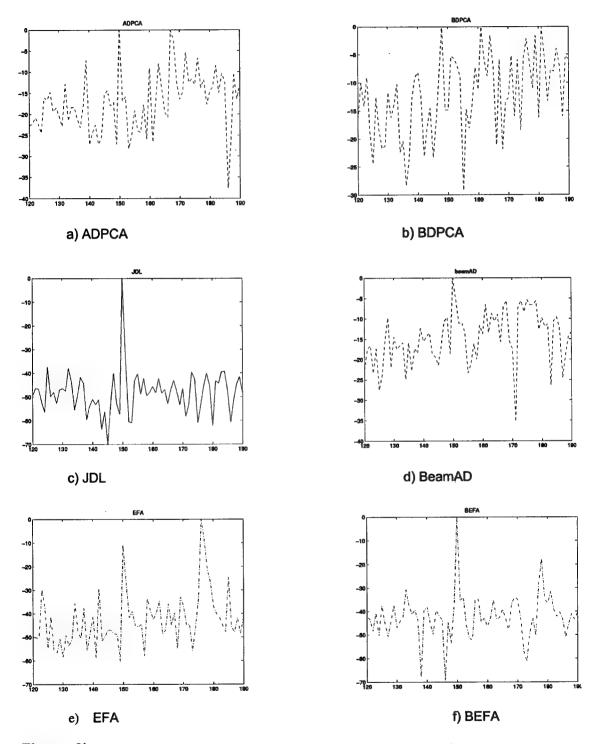


Figure 8) Performance comparison when target is inserted at range 150, case g.

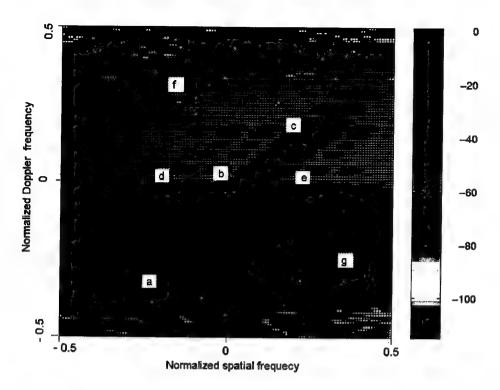


Figure 9) Power spectrum plot of range 350.

Case	Normalized Doppler Frequency	Normalized Spatial Frequency	The 3 best schemes	D .	Figure
а	-0.2656	-0.2656	EFA JDL BEFA	24 18 5	10
b	-0.0312	-0.0312	BEFA EFA JDL	15 12 10	11
С	0.125	0.203	JDL EFA BEFA	34 25 13	12
d	-0.0312	-0.1875	EFA JDL ADPCA	28 20 12	13
е	-0.0312	0.203	EFA JDL ADPCA	22 18 7	14
f	0.3593	-0.1875	EFA JDL BeamAD	48 43 24	15
g	-0.1875	0.3593	JDL BEFA BeamAD	33 18 14	16

Table 3) The three best schemes for all the cases when the target is inserted at range 350. (D is the approximate difference between the normalized test statistic at the target and the largest peak in the normalized test statistic at some other range.)

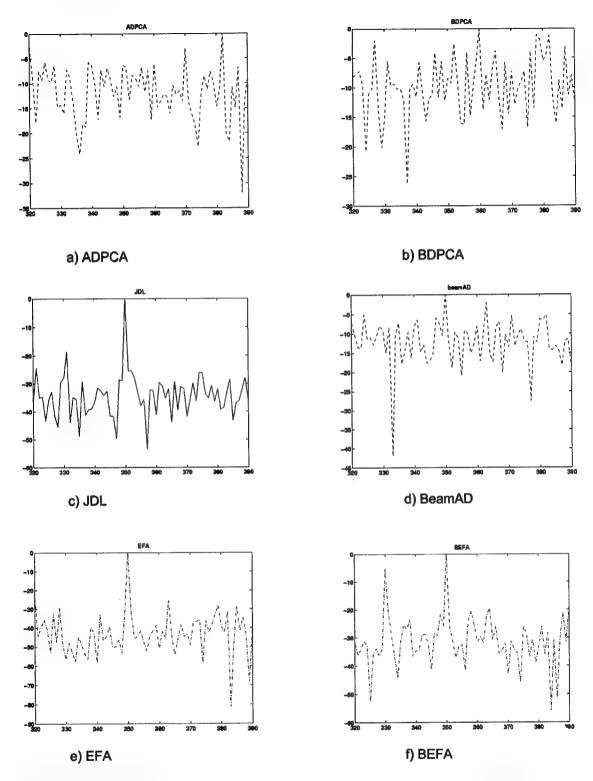


Figure 10) Performance comparison when target is inserted at range 350, case a.

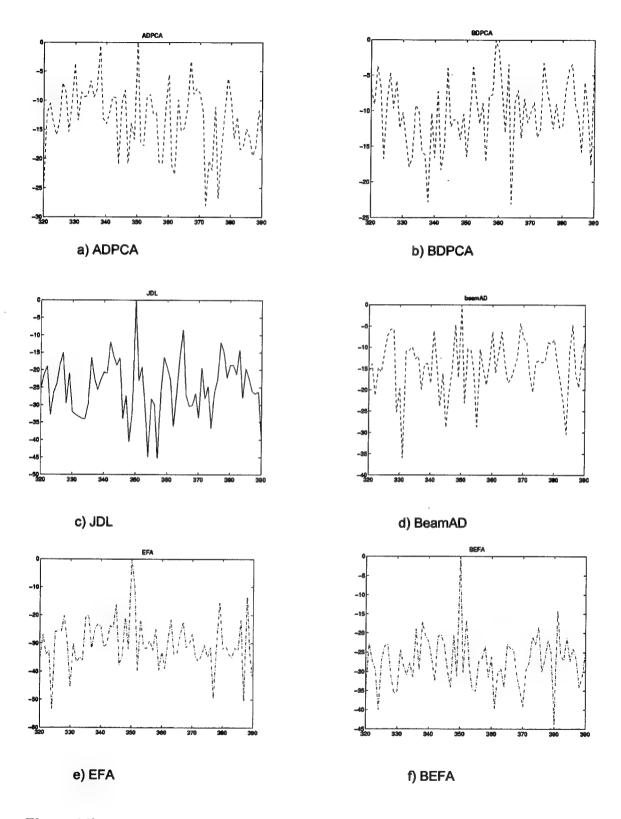


Figure 11) Performance comparison when target is inserted at range 350, case b.

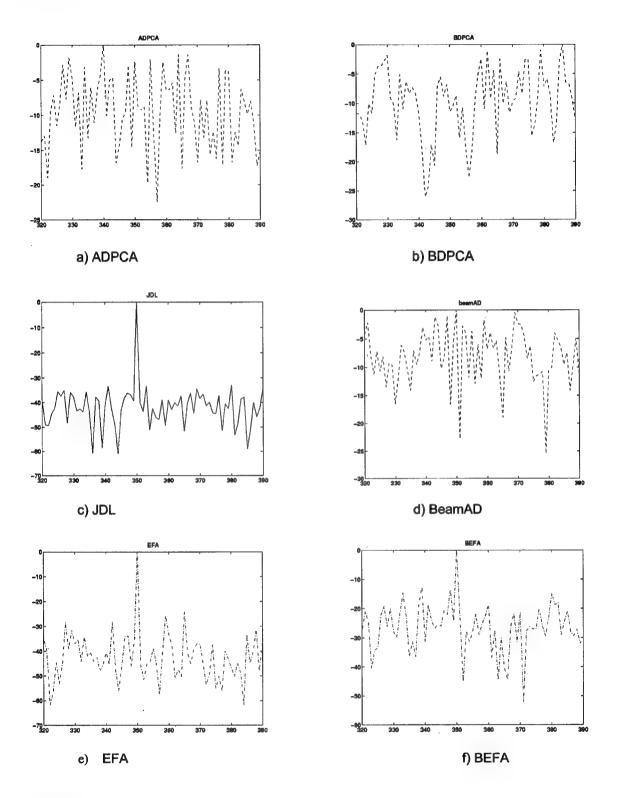


Figure 12) Performance comparison when target is inserted at range 350, case c.

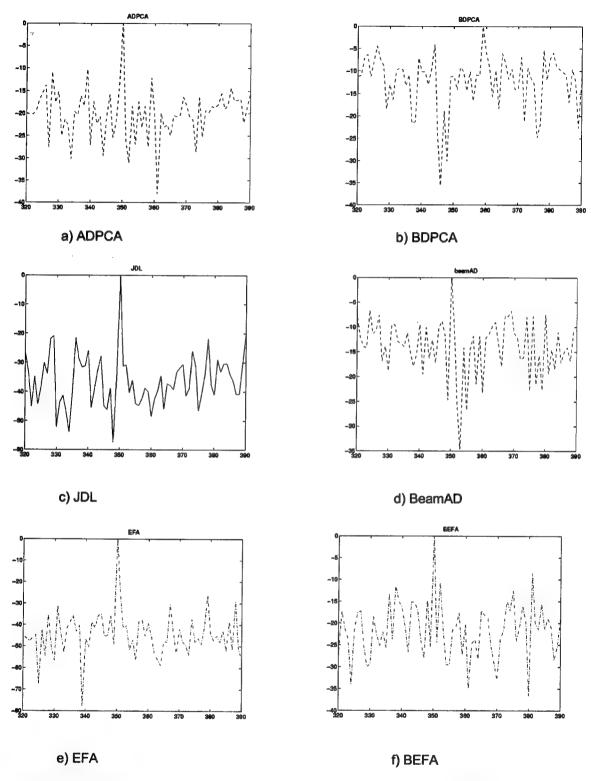


Figure 13) Performance comparison when target is inserted at range 350, case d.

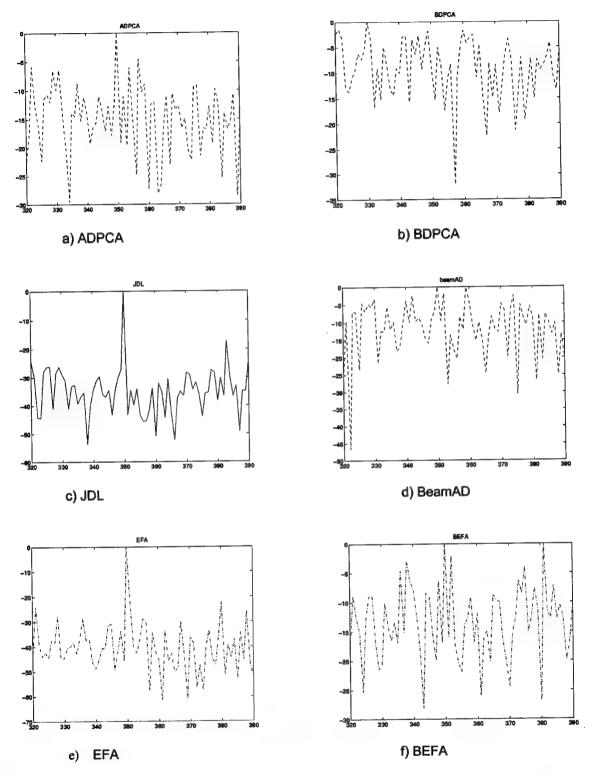


Figure 14) Performance comparison when target is inserted at range 350, case e.

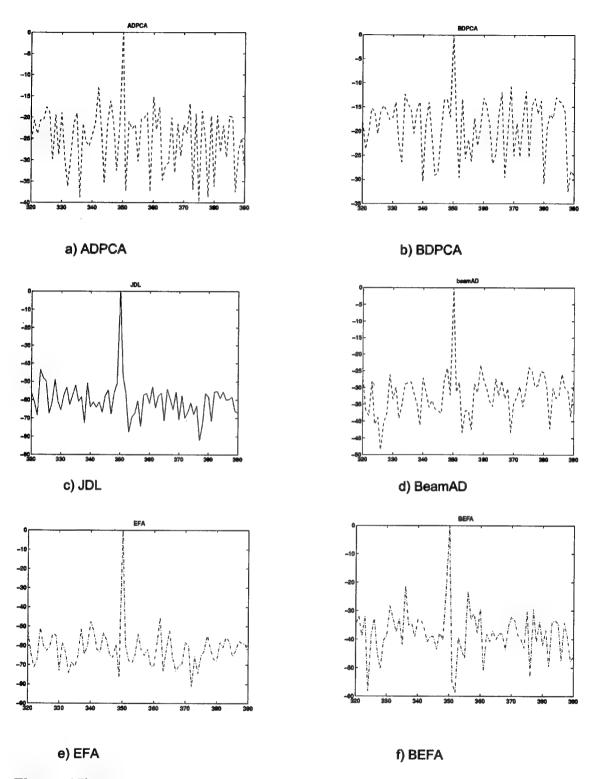


Figure 15) Performance comparison when target is inserted at range 350, case f.

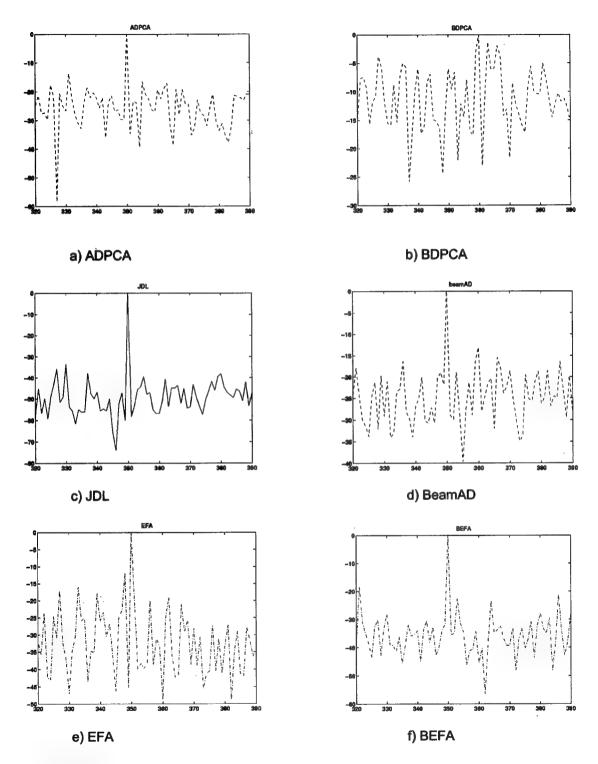


Figure 16) Performance comparison when target is inserted at range 350, case g.

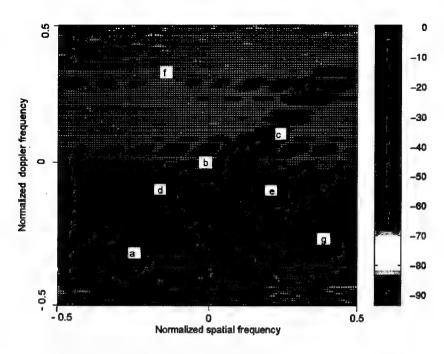


Figure 17) Power spectrum plot of range 415

Case	Normalized Doppler Frequency	Normalized Spatial Frequency	The 3 best schemes	D	Figure
а	-0.2656	-0.2656	EFA JDL ADPCA	25 22 12	18
b	-0.0312	-0.0312	BEFA BeamAD EFA, ADPCA	25 18 15	19
С	0.125	0.203	BEFA JDL BeamAD	30 23 12	20
d	-0.0312	-0.1875	JDL, BEFA BeamAD EFA, ADPCA	20 18 16	21
е	-0.0312	0.203	EFA JDL ADPCA	28 24 15	22
f	0.3593	-0.1875	JDL BeamAD BEFA	45 35 32	23
g	-0.1875	0.3593	JDL EFA BeamAD	42 20 18	24

Table 4) The three best schemes for all the cases when the target is inserted at range 415. (D is the approximate difference between the normalized test statistic at the target and the largest peak in the normalized test statistic at some other range)

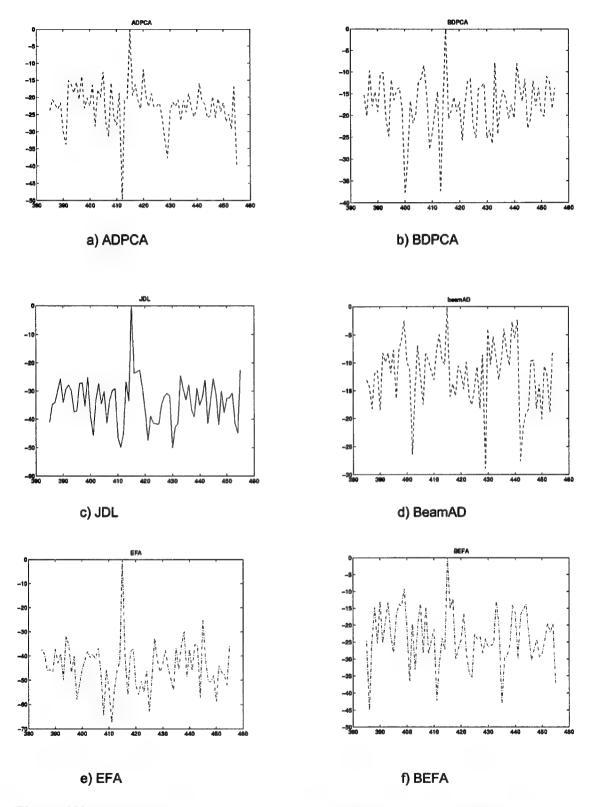


Figure 18) Performance comparison when target is inserted at range 415, case a.

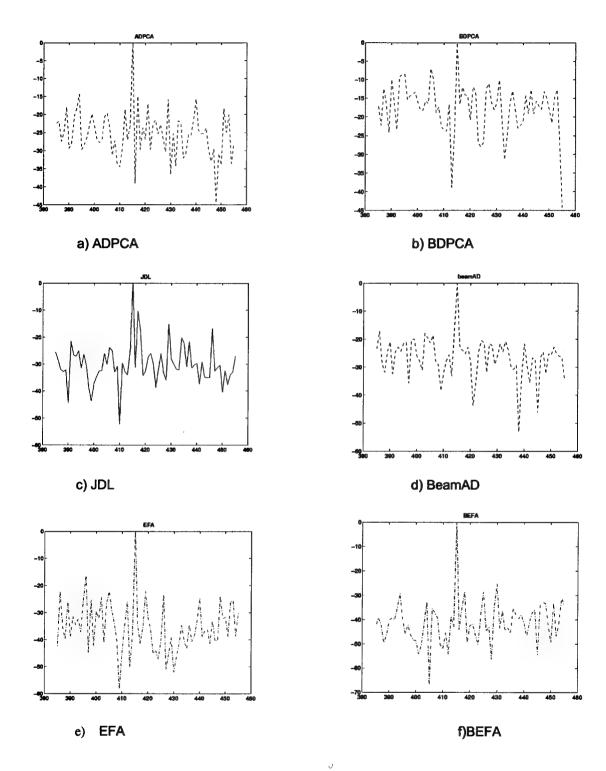


Figure 19) Performance comparison when target is inserted at range 415, case b.

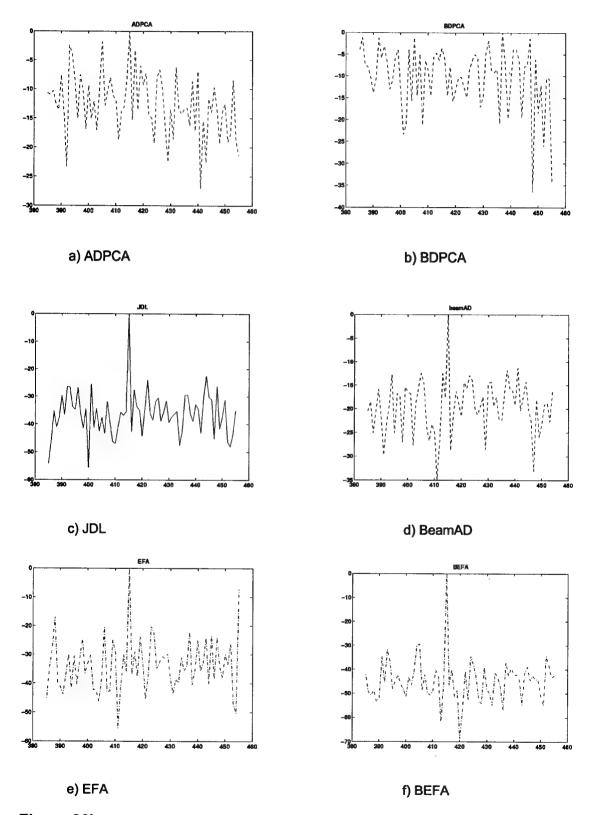


Figure 20) Performance comparison when target is at range 415, case c.

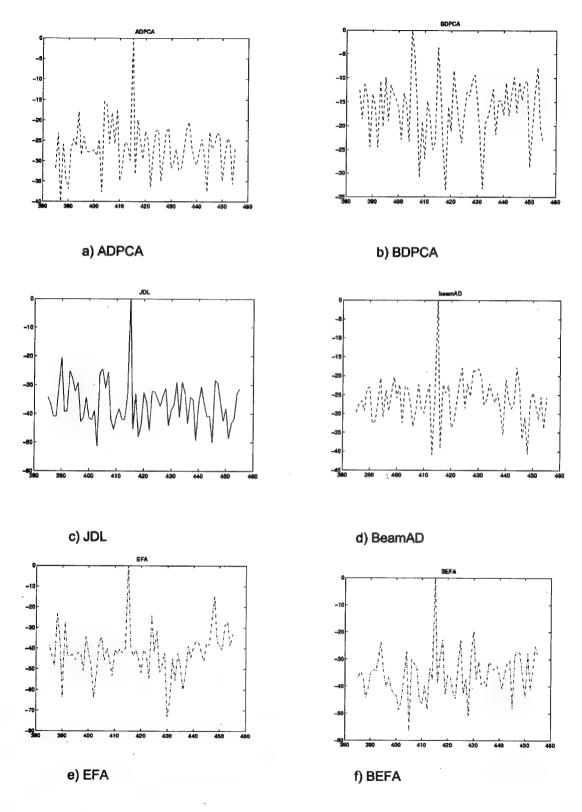


Figure 21) Performance comparison when target is inserted at target range 415, case d.

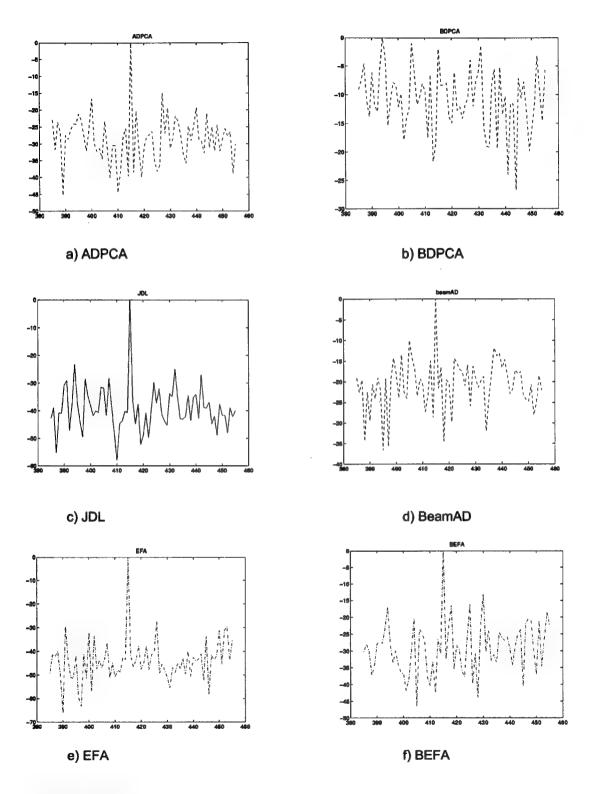


Figure 22) Performance comparison when target is inserted at range 415, case e.

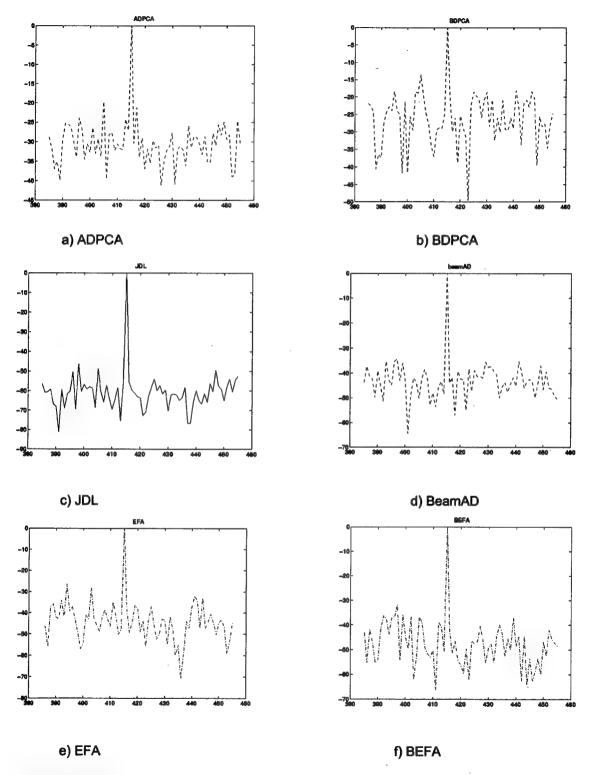


Figure 23) Performance comparison when target is inserted at range 415, case f.

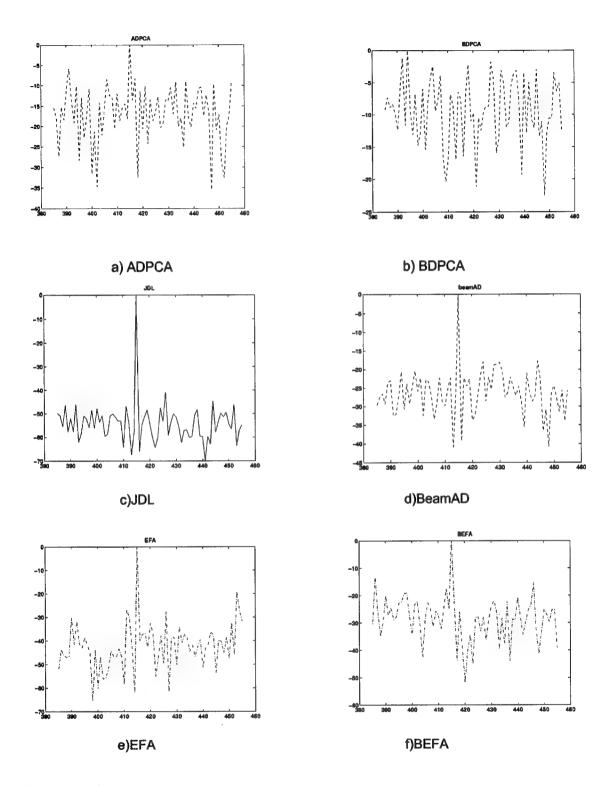


Figure 24) Performance comparison when target is inserted at range 415, case g.

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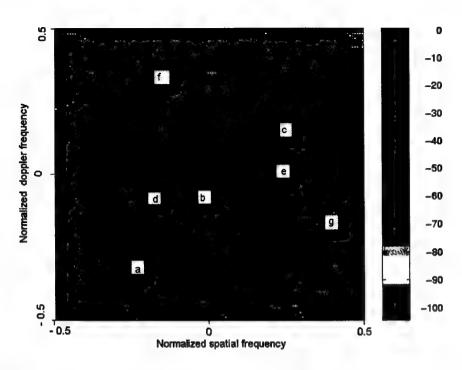


Figure 25) Power spectrum plot of range 500

Case	Normalized Doppler Frequency	Normalized Spatial Frequency	The 3 best schemes	D	Figure
а	- 0.2656	- 0.2656	JDL EFA ADPCA	32 13 7	26
b	- 0.0312	- 0.0312	JDL EFA BEFA	22 20 15	27
С	0.125	0.203	JDL BEFA EFA, BeamAD	36 12 7	28
ď	- 0.0312	- 0.1875	JDL EFA ADPCA	26 20 10	29
е	- 0.0312	0.203	ADPCA JDL EFA	22 18 15	30
f	0.3593	- 0.1875	JDL EFA, BEFA BeamAD	38 28 23	31
g	- 0.1875	0.3593	BEFA, JDL BeamAD EFA	33 17 15	32

Table 5) The three best schemes for all the cases when the target is inserted at range 500. (D is the approximate difference between the normalized test statistic at the target peak and the largest peak in the normalized test statistic at some other range)

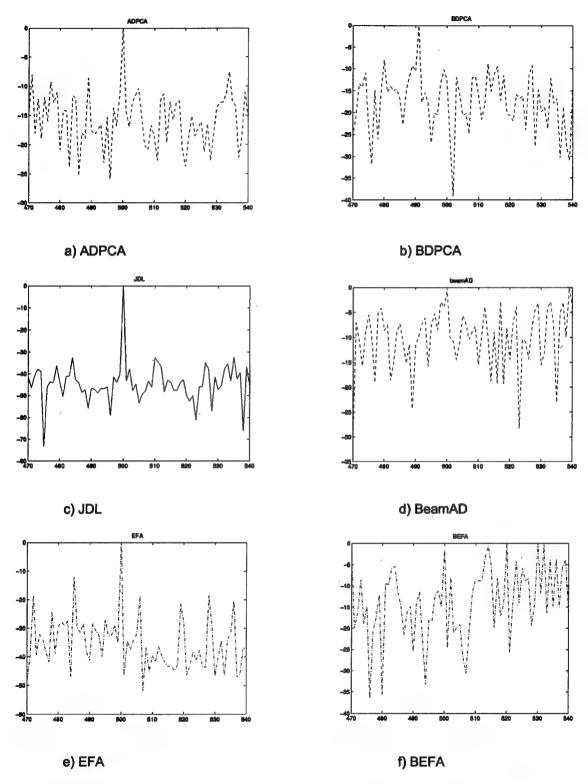


Figure 26) Performance comparison when target is inserted at range 500, case a.

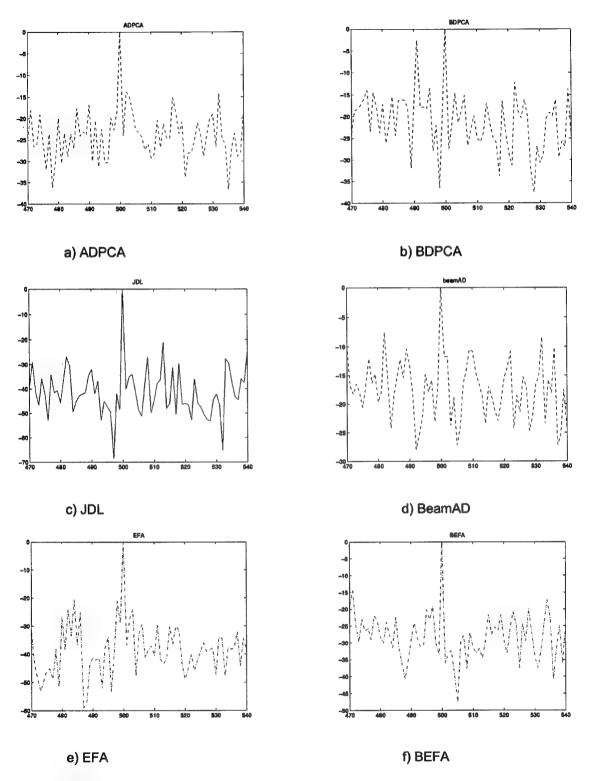


Figure 27) Performance comparison when target is inserted at range 500, case b.

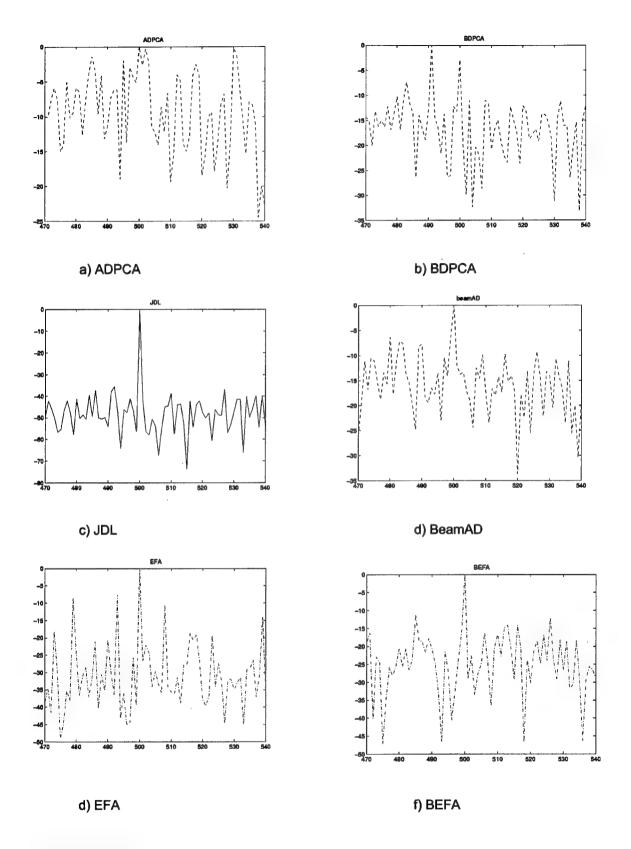


Figure 28) Performance comparison when target is inserted at range 500, case c.

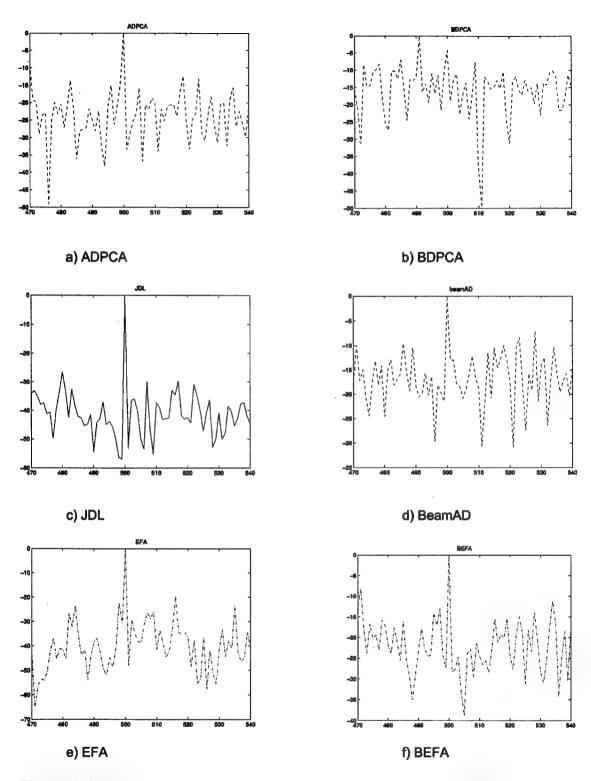


Figure 29) Performance comparison when target is inserted at range 500, case d.

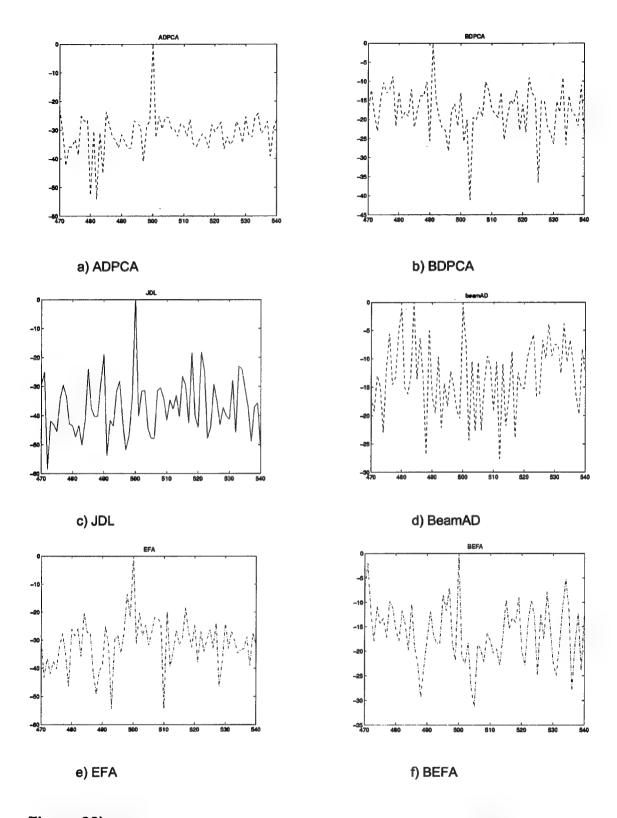


Figure 30) Performance comparison when target is inserted at range 500, case e.

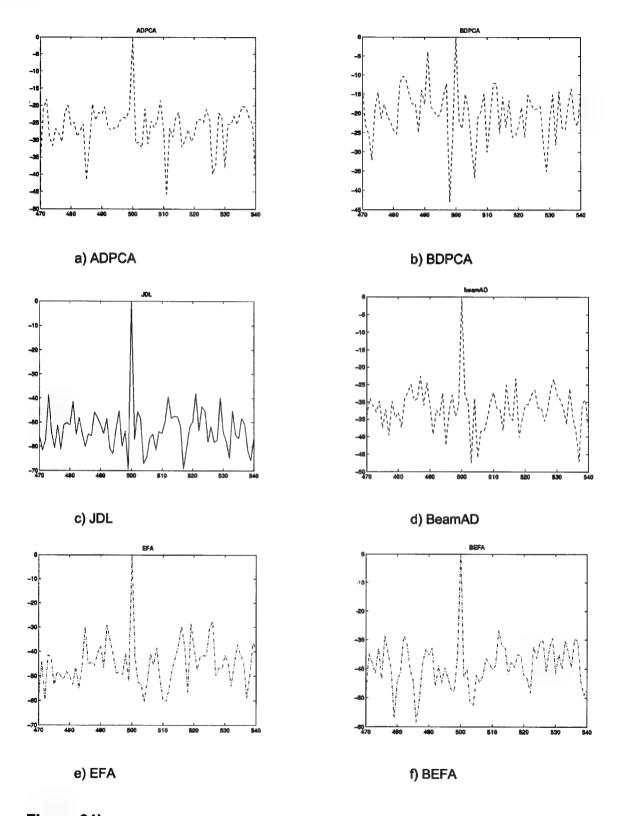


Figure 31) Performance comparison when target is inserted at range 500, case f.

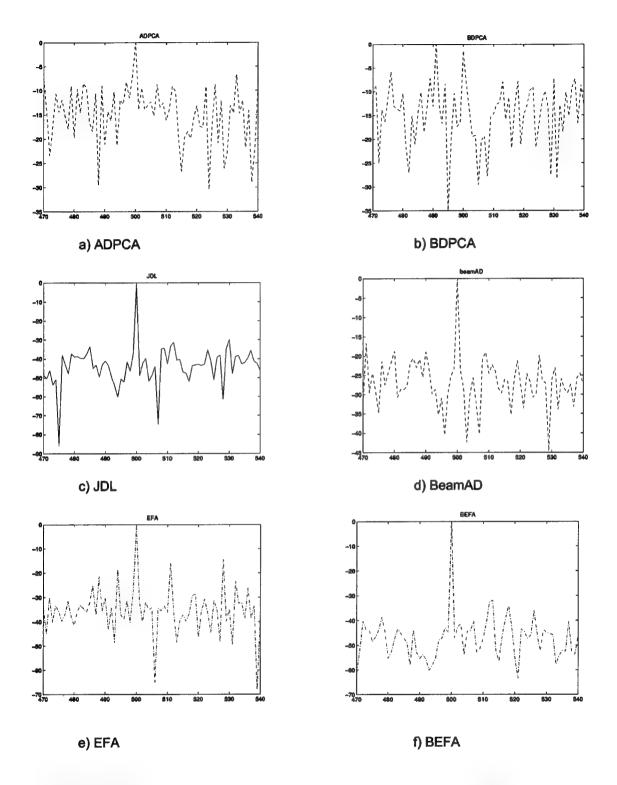


Figure 32) Performance comparison when target is inserted at range 500, case g.

APPENDIX 1

Scheme	M	N	Kt	Ks	Ap	G	Q
ADPCA	128	22	3	-	$[0_{p\times3}, 1_3, 0_{(125-p)\times3}]^T$	l ₂₂	66
EFA	128	22	3	100	[f _{128,Dp-1} ,f _{128,Dp} ,f _{128,Dp+1}]	l ₂₂	66
FTS	128	22	1	-	f _{128,Dp}	122	22
Beamspace	128	22	3	3	$[0_{p\times3}, 1_3, 0_{(125-p)\times3}]^T$	See	10
ADPCA						Note	
				ŀ			
Subarraying	128	22	3	3	$[0_{p\times3}, I_3, 0_{(125-p)\times3}]^T$	*G	10
ADPCA							
Subarraying	128	22	3	3	[f _{128,Dp-1} ,f _{128,Dp} ,f _{128,Dp+1}]	* G	10
EFA							
Subarraying	128	22	3	3	f _{128,Dp}	* G	4
FTS							
JDL	128	22	3	3	[f _{128,Dp-1} ,f _{128,Dp} ,f _{128,Dp+1}]	See	10
						Note	

Table 6) Parameters for comparison tests in this report.

Notes:

Q is used in (9), in this report.

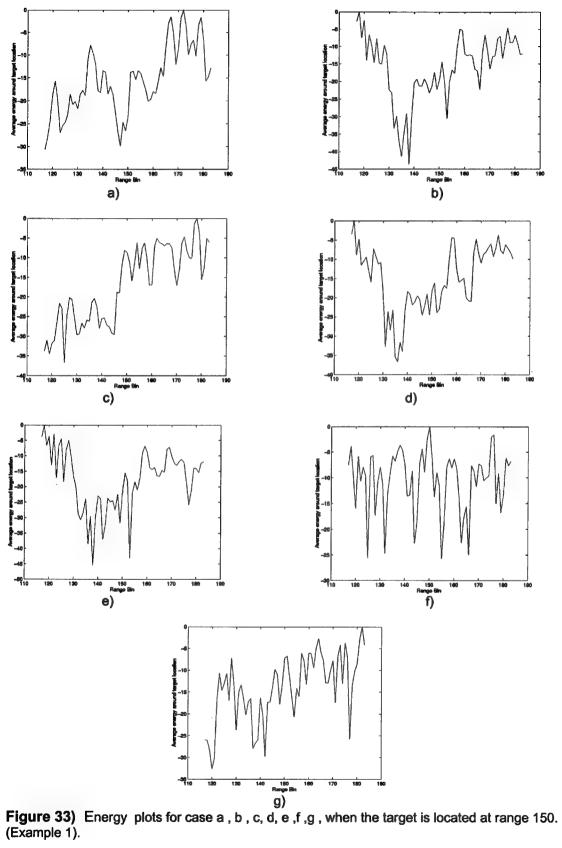
G is the beamforming matrix which is formed as described in (13). When we use the uniform \mathbf{g} vector, the \mathbf{G} ($\mathbf{K_s}$ =3) is as follows

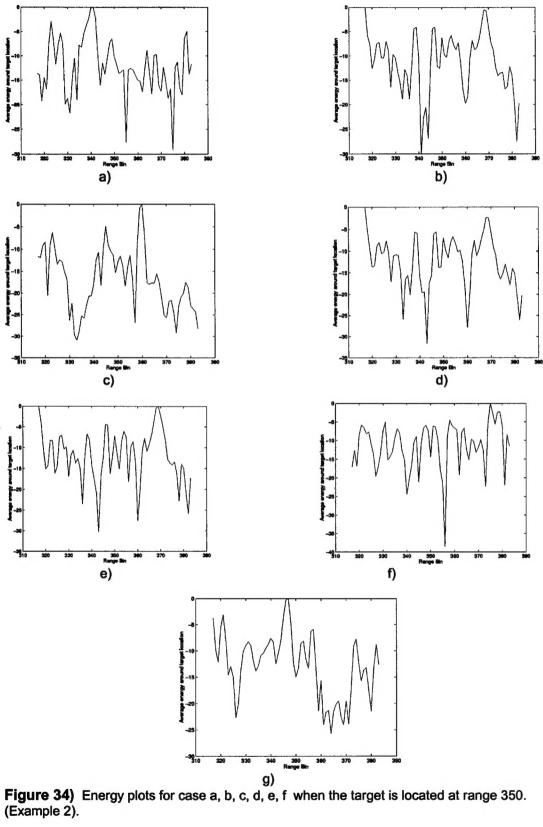
Note: MCARM data is not collected by a uniformly spaced linear antenna array, so its beamforming matrix is relatively complicated. This G matrix used in beamspace ADPCA and JDL can be obtained by using measured steering vector information provided with the MCARM database. See the discussion on modSA at [5].

APPENDIX 2

For real data it is difficult to evaluate if there is a mismatch between the statistics for the reference data and the statistics for the data from the cell-under-test. However by considering energy fluctuations over range, it is possible to partially evaluate this. When the mismatch is near the target location, it could degrade the performance dramatically. To evaluate how stationary the reference data is and to determine how the statistics of the reference data and the statistics for the data from the cell-under-test are mismatched near the target, we plot energy fluctuations over 33 range cells on each side of the target. For each range we plot the average over 11 neighboring Doppler bins and 11 neighboring angle bins near the target Doppler and angle bin. We use 5 Doppler and angle bins on each side of the target. The results are given in Fig. 33 through Fig. 36.

In Fig. 33 through Fig. 36, in all plots the data seems to be non-stationary. But we judge that when the number and the duration of the humps and the nulls increases in these plots, the degree of non-stationary of the data also increases. Generally when the data is highly non-stationary the beamspace post-doppler schemes JDL and BEFA outperform EFA and the other element space schemes.





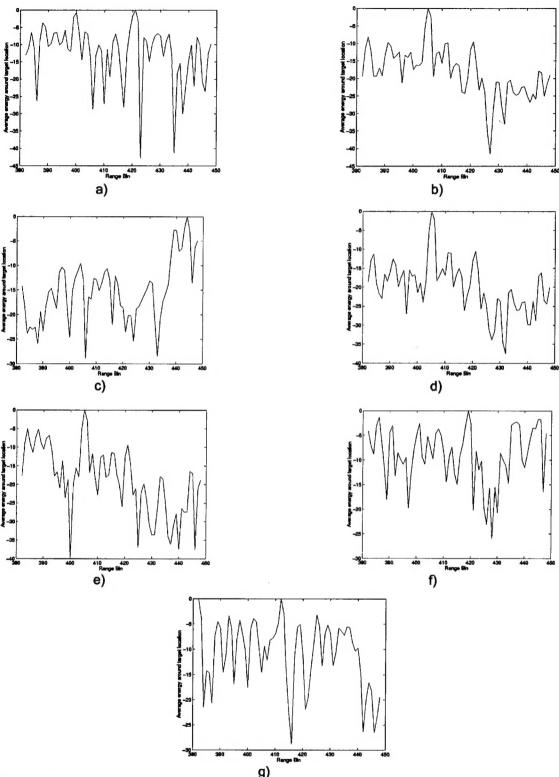


Figure 35) Energy plots for case a, b, c, d, e, f, g when the target is located at range 415. (Example 3).

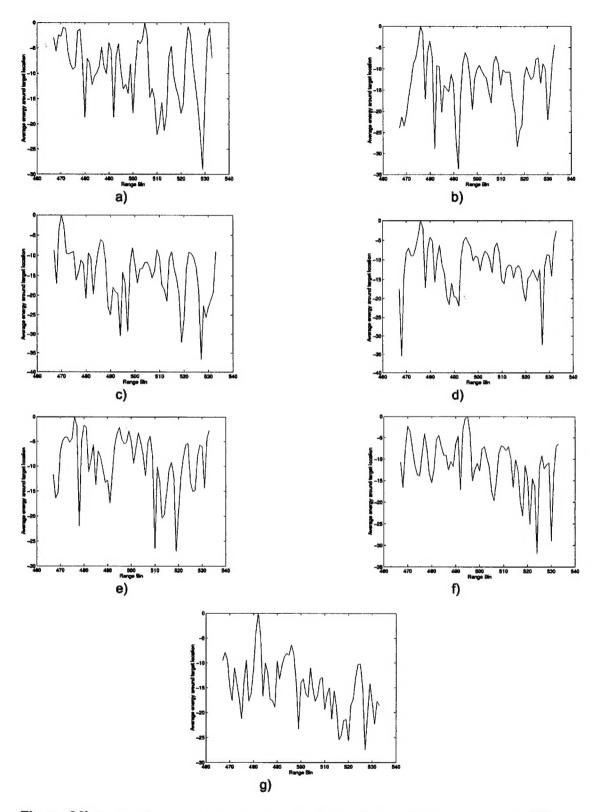


Figure 36) Energy plots for case a, b, c, d, e, f, g f when the target is located at range 500. (Example 4)

4. REFERENCES

- J.Ward, Space-Time Adaptive Processing for Airborne Radar, Technical Report 1015, MIT Lincoln Laboratory, December, 1994.
- Z. Gu, Performance Comparison of STAP Schemes For Airborne Radar, Masters Thesis, Lehigh University, December, 1996.
- 3. A. Oppenheim and R. Schafer, *Discrete-Time Signal processing*, Prentice Hall, Englewood Cliffs, New Jersey, 1989.
- M. D. Zoltowski, "Beamspace M.L. Bearing Estimation for Adaptive Phased Array Radar", In Adaptive Radar Detection and Estimation, S. Haykin and A. Steinhardt Editors, John Wiley and Sons, New York, 1992.
- 5. MCARM database web site, Rome Laboratory USAF, 'http://128.132.2.229'
- * 6. D. Sloper et al., MCARM Final Report, Rome Lab Tech. Rept., RL-TR-96-49, April 1996.
 - 7. L W. Melvin and B. Himed, "Comparative Analysis of Space –Time Adaptive Algorithms with Measured Airborne Data", presented at the 7th International Conference on Signal Processing Applications and Technology, October 7-10, 1996.
 - 8. W. Melvin, H.Wang and M. Wicks, "Multichannel Airborne Radar Array Data Analysis", presented at 41st Annual Tri-Service radar symposium, June 1995.
 - 9. H. Wang, "Space-Time Processing and Its Radar Applications", notes used for the class ELE 891 at Syracuse University, Summer 1995.

^{*}Although this report references the above limited report, no limited information has been extracted.